

Largely based on Jonathan Gair's Lecture notes and Maggiore's book!

(Signal processing!)

$$s(t) = u(t) + n(t)$$

\uparrow \uparrow
signal noise



How to find the signal?

Noise $n(t)$ evolves according to probability laws.

$$p_N(u_N(t_N); \dots; u_1, t_1) \propto e^{-\frac{1}{2} \int_{t_1}^{t_N} \dot{u}_i^2 dt_i}$$

Stationarity of noise: Only depends on time difference

$$p_N(u_N, t_N; \dots; u_1, t_1) = p_N(u_N, t_N; \dots; u_1, t_1) \Big|_{T - \Delta t}$$

Can model noise as Gaussian

$$p \sim = A \exp \left[-\frac{1}{2} \sum_j \alpha_j \int_{t_1}^{t_N} (y_j - \bar{y})(y_j - \bar{y}) \right]$$

$\bar{y} = g(t, u_i)$

Zero noise:

$$\int_{-\frac{T}{2}}^{\frac{T}{2}} u(t) dt = 0$$

How about fluctuations? $\tilde{P} = \int_{-\frac{T}{2}}^{\frac{T}{2}} |u(t)|^2 dt$

Power

$\sim \text{const} \rightarrow \bar{P} = \text{const} \times T \text{ linear}$

$$\Rightarrow P_n = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-\frac{T}{2}}^{\frac{T}{2}} |\hat{u}(t)|^2 dt$$

Remember $\int_{-\infty}^{\infty} f(x) dx = \int_{-\infty}^{\infty} f(f) df$ Parseval.

Noise Real $\Rightarrow \int_{-\infty}^{\infty} |\hat{u}(t)|^2 dt = 2 \int_0^{\infty} |\hat{u}_r(f)|^2 df$

$\hat{u} \propto \sqrt{\sum}$

$$\Rightarrow P_n = \lim_{T \rightarrow \infty} \frac{2}{T} \int_0^{\infty} |\hat{u}_r(f)|^2 df$$

↑ only f

define $= \underbrace{\int_0^{\infty} \lim_{T \rightarrow \infty} \frac{2}{T} |\hat{u}_r(f)|^2 df}_{\text{S}_n}$

S_n spectral density

$$= \int_0^{\infty} S_n(f) df$$

From the definition: Variation around Af

$$\Delta n^2 = \lim_{N \rightarrow \infty} \frac{2}{N} \sum_{n=-\frac{N}{2}}^{\frac{N}{2}} \left| \frac{1}{\Delta r} \int_{n\Delta r}^{(n+1)\Delta r} u(t) e^{2\pi ift} dt \right|^2 = \frac{S_n(f)}{\Delta r}$$

$\simeq S_n(f) \Delta f$

$$\Rightarrow \boxed{\Delta u_{\text{avg}} = \int S(f) \Delta f}$$

Wiener-Khintchine

Auto-correlation of u is related to the spectral density

$$\boxed{\langle u^*(t) u(t+\tau) \rangle = \int S_u(f) e^{j2\pi f \tau} df}$$

$$\lim_{T \rightarrow \infty} \frac{1}{T} \int_{-\frac{T}{2}}^{\frac{T}{2}} u^*(t) u(t+\tau) dt$$

Fourier transform

$$\boxed{\langle \hat{u}^*(f) \hat{u}(f') \rangle = S_u(f) \delta(f-f')}$$

Different types of noise

$$S_u(f) = \text{const} \quad \text{white}$$

$$S_u(f) = \frac{1}{f} \quad \text{flitter}$$

$$S_u(f) = \frac{1}{f^2} \quad \text{random walk}$$

$$\bar{P}_h = \frac{1}{\Delta t} \int_0^{\Delta t} |h(t)|^2 dt = h_c^2$$

Burst

↑
characteristic amplitude

bandwidth Δf
duration Δt
frequency f

From above noise power

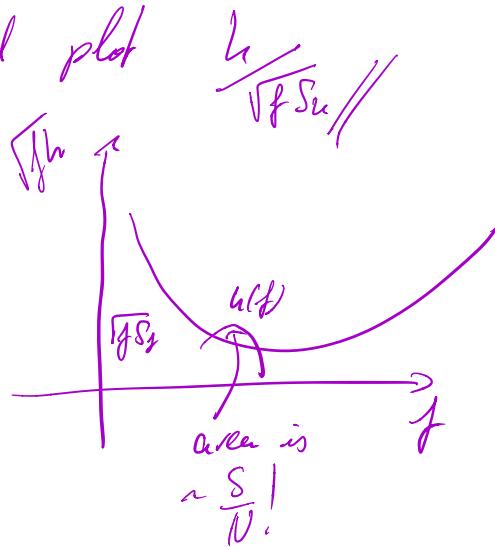
$$\int_{-\infty}^{\infty} P_{\text{noise}} \, d\nu$$

$$|u|^2 \sim \Delta f S_u(f)$$

$$\Rightarrow \left(\frac{s}{N}\right)^2 = \frac{\bar{P}_u}{\Delta f S_u(f)} = \frac{h c^2}{\Delta f S_u(f)} = \frac{h c^2}{f S_u(f)}$$

broad band shot. $\Delta f \approx f$

For a burst we should plot



Alternative (LST band)

continuous CW Frequency largely constant

$$h(t) = h_0 e^{2\pi i f_0 t}$$

$$\Rightarrow P_u = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} |h(t)|^2 dt = \frac{1}{2} h_0^2$$

If f_0 const then error

$$\boxed{\Delta f \sim \frac{1}{T}}$$

$$\Rightarrow |u|^2 \sim \Delta f S_u \sim \frac{S_u}{T}$$

So noise Lee goes like $\sqrt{\frac{S_u}{T}}$
 ↗
 observation time

What about position dependent effects?

$$\langle S_f(t) \rangle \approx S_{f_f}(t) \quad \text{LIGO}$$

↓
by position
averaged

An instead of compact binaries has finite time
 and energy $\rightarrow \frac{1}{T} \bar{h}(t) \rightarrow 0 \quad T \rightarrow \infty$

Need to windows the above expression

$$w(t) = \int_{-\infty}^{\infty} K(t-t') s(t') dt'$$

↑
windowing kernel

$$\left(\frac{S}{N} \right)(t) = \frac{\int K(t-t') h(t') dt'}{\sqrt{\int K(t-t') w(t') dt'}}$$

↑
induced noise

Remember

$$\boxed{s = h + n}$$

$$\rightarrow \langle s \rangle_f = \langle h \rangle_f + \frac{\langle h \rangle_f}{\left(\frac{S}{N} \right)}$$

want to maximize this

In Fourier space $\langle u(t) \hat{u}(t') \rangle = \hat{U}^* \hat{U}$

$$\Rightarrow \frac{S}{N} = \frac{\int \hat{K} \hat{U} \hat{U}^* dt}{\int |\hat{K}(f)|^2 S_u(f) df}$$

$\hat{U}^* \hat{U} \sim \int S_u^2 \sim \int S_u$

Define optimal filter

$$(h_1 | h_2) = 2 \int_0^\infty \frac{\hat{h}_1^* h_2 + \hat{h}_1 \hat{h}_2^*}{S_u(f)} df$$

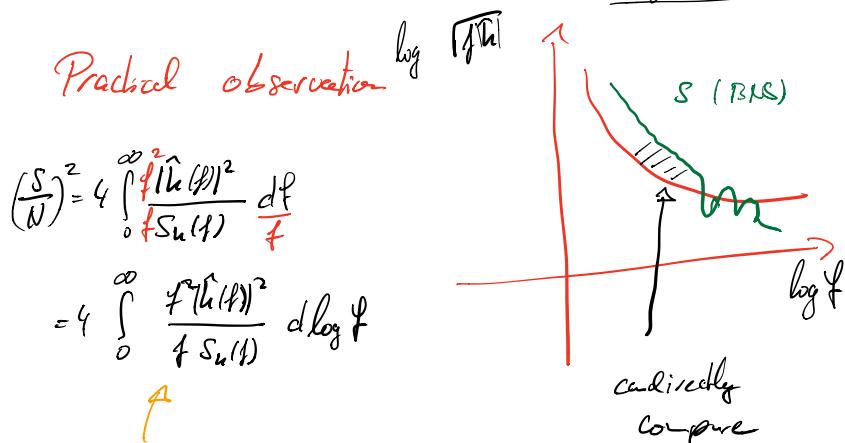
$$\Rightarrow \frac{S}{N} = \frac{(S_u K | h)}{\int (S_u K | S_u K)}$$

Maximize
linear
filter
 $\hat{K} \propto \frac{\hat{h}_{\text{optimal}}}{S_u}$

if $h = h_{\text{optimal}}$
 $\Rightarrow \left(\frac{S}{N}\right)[h] = \sqrt{h(h)}$ contains S_u !!

Standard Signal to Noise

Practical observation



compare log curves

$$\log \left(\frac{f^2 |\hat{h}(f)|^2}{f S_u(f)} \right) = 2 \log (\sqrt{f h}) - 2 \log (\sqrt{f S_u})$$

$$0 \cdot \text{f}_{\text{obs}}(t) / 0 \cdot \text{f}_{\text{true}}(t)$$

$$\text{Area in log-log plot} \propto \frac{S}{N}$$

Parameter estimation

Template will depend on $\theta_1, \dots, \theta_D$

$$R \approx \frac{\hat{L}(f, \theta_m)}{S_n(f)}$$

From Wiener-Khinchine

$$\rightarrow \langle u^* u \rangle = \frac{1}{2} S_u(f) \delta(f-f')$$

$$\Rightarrow P(u_0) = N e^{-\frac{1}{2} \int_{-\infty}^{\infty} \frac{|u_0(\ell)|^2}{S_u(\ell)} d\ell}$$

↑
gaussian noise

$$= N e^{-\frac{(u_0)^2}{2}}$$

$$\Rightarrow \text{assume } s(t) = L(t, \theta_s) + u_s(t)$$

How likely is the signal?

$$p(u_0) = N e^{-\frac{(s-u)^2}{2}} =: \Lambda(s, \theta_t)$$

$$\Rightarrow \Lambda = N e^{(u(s) - \frac{1}{2}(s/s))} e^{-\frac{1}{2}(u|u)}$$

\uparrow
prior probability p^0

$\Lambda_s = N e^{(u(s) - \frac{1}{2}(u|u))}$

$$\rightarrow \text{Bayes } P(\mathcal{D}|s) = P(\theta_s) P(s|\mathcal{D}) \\ = P(\mathcal{D}_s) \Lambda_s(s|\mathcal{D})$$

Need to fit the parameters:

For flat prior probability maximum likelihood estimate

$$\log p \approx \log 1 = (l/s) - \frac{1}{2} (l/l) + \text{const}$$

$$\text{extremize } \log p \Rightarrow \underbrace{(D_i l / s) - (D_i l / l)}_{\uparrow k \text{ update}} = 0$$

This will find some local maximum

Bayes estimator

$$D_B^i(s) = \int d\theta D^i p(\theta | s) \quad \leftarrow \text{Bayesian estimate}$$

$$\Sigma_B^{ij} = \int d\theta [D^i - D_B^i] [D^j - D_B^j] p(\theta | s) \quad \leftarrow \text{for "optimal" error}$$

mean square deviations.

$$\text{Assume that we have a good guess } D_B^i = D^i - \Delta D^i$$

$$\Rightarrow p(\theta | s) = \sqrt{\frac{1}{2} \Gamma_{ij} \Delta D^i \Delta D^j} \quad \text{taylor expand in } \Delta D^i$$

$$\Gamma_{ij} = (D_i D_j l / l - s) + (D_i l / D_j l)$$

$$\text{for high } S_N \rightarrow \Gamma_{ij} \approx (D_i l / D_j l)$$

$$\langle \Delta D^i \Delta D_j \rangle \approx (\Gamma^{-1})^{ij} \quad \text{property of the template:}$$

How to beat QO signals of inspiralling
binaries?

the simple template

$$h(t) = A_r (\pi f_c)^{2/3} \cos(\Phi(f_{c0}) + \Phi_0)$$

$$+ A_x (\pi f_{c0})^{2/3} \sin(\Phi(f_{c0}) + \Phi_0)$$

$$A_r = \frac{4}{r} (M_c)^{5/3} F_r(\vartheta, \phi) \frac{1 + \cos^2 \alpha}{2}$$

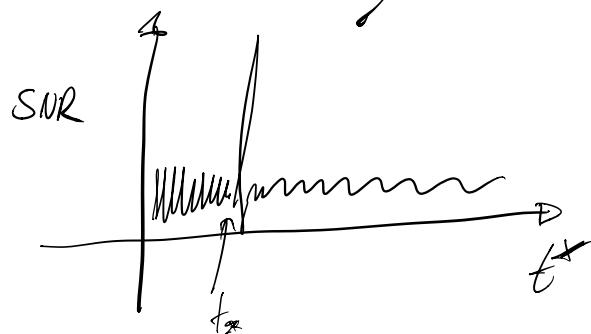
need
to fix Φ_0

$$A_x = \frac{4}{r} M_c^{5/3} F_x(\vartheta, \phi) \cos \alpha$$

Determining temporal offset

just appears as
a phase
in Fourier
space

$$(h(\vartheta, t_*)|s) \stackrel{\text{real}}{\rightarrow} \text{Re} \int_0^\infty df \frac{\hat{h}^*(f, \vartheta) \hat{s}(f)}{S_h(f)} e^{i 2\pi f t_*}$$



How to fix Φ_0

$$h(t) = h_c(t) \cos \varphi + h_s(t) \sin \varphi \quad | \text{real space}$$

Let's use the maximum likelihood estimator to get A

$$\text{Assume } h = Ah_A(s, \theta_m)$$

$$\Rightarrow \log \Lambda(s, \theta_m) = A(h_A(s)) - \frac{A^2}{2}(h_A(h_A))$$

$$\frac{\partial \log \Lambda}{\partial \theta_m} = 0 \Rightarrow \boxed{A = \frac{h_A(s)}{h_A(h_A)}} \quad \text{fixes the constant}$$

$$\Rightarrow \boxed{\log \Lambda(s, \theta_m) = \frac{1}{2} \frac{(h_A(s))^2}{h_A(h_A)}}$$

$$\Rightarrow 2 \log \Lambda = \frac{(h(s))^2}{(h/h)} = \frac{[(h_A(s) + (h_s(s) \tan e)]^2}{(h_A(h_A) + (h_s(h_A) \tan^2 e + 2(h_A(h_A) \tan e))}$$

Can introduce $\Phi_p, \Phi_q \Rightarrow$

$$h_p = h_c \cos \varphi_p + h_s \sin \varphi_p$$

$$h_q = h_c \cos \varphi_q + h_s \sin \varphi_q$$

$$\Rightarrow (h_p/h_q) = 0$$

$$\Rightarrow 2 \log \Lambda = \frac{(h_p(s))^2}{(h_p/h_p)} + \frac{(h_q(s))^2}{(h_q/h_q)}$$



Can maximize these

On the detectability of sources:

$$\hat{h}(f) = \sqrt{\frac{S}{6}} \cdot \frac{1}{2\pi} \cdot \frac{M_c^{5/6}}{D} \cdot f^{-7/6} e^{i\phi} Q(\omega, \theta, z)$$

orientation

Using the most general matched filter

$$\Rightarrow \left(\frac{S}{N}\right)^2 = 4 \int_0^{\infty} df \frac{|\hat{h}(f)|^2}{S_n(f)}$$

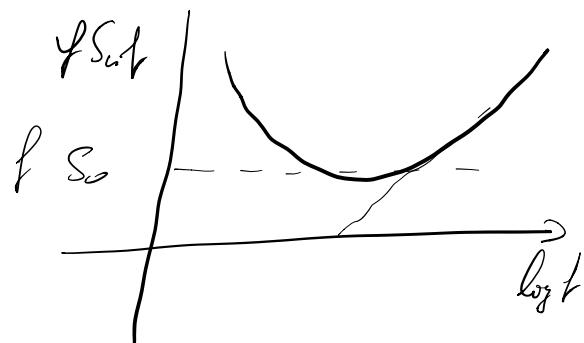
from

$$\left(\frac{S}{N}\right)^2 = \frac{S}{6} \frac{1}{\pi^{5/3}} \frac{M_c^{5/3}}{D^2} |Q|^2 \int_0^{\infty} df \frac{f^{-7/3}}{S_n(f)}$$

angle average $\langle |Q| \rangle \approx \frac{2}{5}$

$$D \sim \frac{M_c^{5/6}}{\left(\frac{S}{N}\right)} \sqrt{\int_0^{\infty} df \frac{f^{-7/3}}{S_n(f)}}$$

detectable distance
scales $\frac{1}{\left(\frac{S}{N}\right)}$



Let's do some simple assumptions:

$$S \approx S_0 = \text{const}$$

$$\Rightarrow \frac{S}{N} \sim \frac{1}{D} M_c^{5/6} S_0^{-\frac{1}{2}} f_0^{-\frac{5}{3}}$$

Number of cycles: $N_c = M_c^{-5/3} f_0^{-5/3}$

$$h_0 \sim \frac{1}{D} f_0^{2/3} M_c^{5/3} \quad (\text{real space!})$$

$$\Rightarrow \frac{S}{N} \sim \frac{h_0}{\sqrt{f_0 S_0}} N_c^{1/2} \quad \text{if } \frac{S}{N} \text{ improves with number of orbits}$$

$\frac{S}{N}$ of a single burst